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News Features

Probing the News

129 China's three-way stretch
137 Brokers put stock in electronics

Electronics Review

37 Communications: The chirp approach
37 Avionics: With flying colors
38 Oceanography: Staying put
39 Components: Strengthening a link . . . ; or replacing it
40 Military electronics: Where are (were) we?; Repairs needed; Vastly important
42 Solid state: So who's worried?
44 Advanced technology: Strangest places
46 Computers: Long and short of it
47 For the record

Electronics Abroad

187 International: Economic slowdown
188 West Germany: The right type
188 Japan: Fit to print; IC's where it counts
190 Great Britain: High on ions
191 France: Fast film

New Products

147 Low threshold voltages switch MOS registers
150 New components review
150 New components: Power transistors— the jump to 50 watts
154 New instruments review
154 New instruments: Off-the-shelf IC's in new counter; Tiny power supply puts out 17.5 kv
159 New subassembly review
159 New subassemblies: Increasing accuracy of telemetered data
163 New microwave review
163 New microwaves: Mesa structure quiets noise
167 New semiconductor review
167 New semiconductors: Integrating a silicon controlled rectifier; Fine tuner for tv fits on a chip

Technical Articles

I. Design

Integrated electronics 78 Linear IC's, part 2
78 Heart of the matter
Differential amplifier is the basic configuration
John J. Robertson, Motorola Semiconductor Products Inc.
81 Power grab by linear IC's
A monolithic device can drive big power controllers
F. William Gutzwiller and James H. Galloway,
General Electric Co.

Circuit design 87 Designer's casebook
- Sequential switching enables low-frequency multiplication
- Bridge circuit cuts contacts in series-parallel network
- Low temperature triggers diode relaxation oscillator

Circuit design 93 Choking up on LC filters
One choke contains multiple windings instead of many choke filters
Allan G. Lloyd, Avion Electronics Inc.

II. Applications

Military electronics 99 Designing for the worst of worst cases—nuclear war (cover)
Circuits are hardened to withstand radiation so electronic systems can weather an atomic holocaust
C.F. Johnson

Military electronics 111 Down-to-earth Army antenna
A vertical loop antenna performs as well as a dipole
Kenneth H. Patterson, Army Limited War Laboratory

Departments

4 Readers Comment 25 Electronics Newsletter
8 People 59 Washington Newsletter
14 Meetings 174 New Books
16 Meeting Preview 176 Technical Abstracts
23 Editorial 180 New Literature
185 Newsletter from Abroad

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By Kenneth H. Patterson

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**Fight fans** say a good big man will beat a good little man every time. But Army signalmen setting up radio sites in the jungles of Vietnam would much rather use a small radio antenna with adequate efficiency than a larger, more efficient one that has to be strung on high masts. Mobility and concealment are high-priority features in a guerrilla war.

For communications at medium and high frequencies, the Army normally uses various forms of long wires, dipoles, and rhombics. To be reasonably effective, the long wire should be a minimum of about 100 feet (plus a good ground), the dipole should be at least twice as long, and the better rhombics can exceed 600 feet. In addition, all of these antennas should be elevated to a minimum height of about 40 feet, with 80 feet being better. Such space and height requirements are invariably so difficult to meet that compromises are usually made in the field.

But a solution has been developed at the Army Limited War Laboratory at Aberdeen Proving Ground, Md. A vertical loop antenna (in the shape of an equilateral octagon, 5 feet to a side) has been built that doesn’t have to be raised above the ground. Despite the inherently low radiation resistance of a loop, the new design usually does as good a job as a full-length dipole 40 feet above the ground, even though the length of the dipole is about 234 feet at 2 Mhz and 94 feet at 5 Mhz. One can get an idea of the difference between the low-height performance of vertical loops and horizontal dipoles by visualizing their basic characteristics. The most significant consideration is the effect on the signal of reflection from the ground.

The theory of image antennas tells us that vertically polarized signals are reflected by the earth without a phase shift. Horizontally polarized signals, on the other hand, are phase inverted upon being reflected. The exact amount of inversion depends on orientation, but in most instances it can be expected to approximate the full 180°. It then follows that when the dipole antenna is used at heights lower than about 12λ, the reflected wave has a cancelling effect on the incident wave. In fact, zero height and a perfectly conducting earth are conditions that cause total cancellation and zero radiation.

**Hit the dirt**

Under the same circumstances, however, the reflected wave from a vertical loop combines with the incident wave under what is predominantly an in-phase condition. This quite naturally results in an increase in the amount of energy radiated in useful directions. With greater proximity to the earth, therefore, the radiation from the loop may actually improve. It cannot, of course, exceed a limit of twice its radiation in free space.

A close examination of the expressions for the radiation resistance values of the two antenna types, including the height-modifying terms, shows the loop approaching twice its free-space value as a function of decreasing height, while the horizontal antenna goes to zero.

If we wish to operate our antenna at a really low height, therefore, we must use vertical polarization. And if we cannot accept a null in our overhead
pattern, we are restricted to a vertical loop. The
problem is to improve the efficiency of the loop
antenna.
Besides solving spatial and height problems, the
antenna developed at Aberdeen:

- is operational from below 2.5 Mhz to above 5.5
  Mhz—although tuning ranges as great as 10:1 are
  possible;
- offers a pattern factor well suited to both short-
  and long-range ionospheric propagation with no
  overhead nulls;
- has a predicted efficiency level of from 20% to
  80% throughout the operational band. (Future
  models will do much better.);
- is self-supporting in use;
- can be transported by a small vehicle when
  packed;
- can be set up or dismantled by a crew of no
  more than three men in less than 30 minutes;
- doesn't require an artificial ground plane;
- has sufficient strength to withstand normal
  wind and rain storms.

The power rating, determined by the voltage
breakdown and current ratings of the capacitors
used in the matching network, is arbitrary. A 1-kilo-
wat version has been built.

The loop isn’t being suggested as the best design
for every antenna application. It won’t, for instance,
outperform the large rhombic in specific low-angle
unidirectional tasks. But the loop can do the job
at installations where real estate is limited and a
complex of high antenna masts is impractical.

Field tested
Loop antennas, some shaped as squares and
others as octagons, have undergone extensive field
trials in the U.S. and Vietnam. The results in all
cases have been excellent.

Before the loop was selected, other configura-
tions were considered and rejected. Grounded ver-
ticals, or whips, were ruled out because of their
height and their inherently restricted radiation
pattern.

The horizontal series-fed antennas, including
dipoles, long wires, V’s, and rhombics, were also
dismissed because of their height requirements.
As noted before, the radiation resistance of these
antennas approaches zero as height is reduced.
(Radiation resistance is obtained by dividing the
total radiated power of an antenna by the square
of the effective antenna current measured at the
point where power is supplied.)

The vertical loop, being compact and vertically
polarized, and having no overhead null in its pat-
tern, got the nod almost by default. The superior
low-height performance of a loop antenna was
suggested to the Proving Ground by David Sun-

To improve the loop’s efficiency, two problems
had to be solved: radiation resistance (R_L) had to
be increased, and the sum of all other losses (R_L)
had to be reduced. To assess the problems and
the expected efficiency (E), the standard efficiency
equation was used—efficiency equals radiation re-
sistance divided by the sum of radiation resistance
and R_L.

Cutting losses
The two principal sources of loss in a loop
antenna are the inherent resistance in the con-
ductor used to form the loop and the resistance
in the antenna-matching network.

In considering the losses in the matching net-
work, it was decided the antenna would be driven
from a 50-ohm or 70-ohm coaxial line, thus requir-
ing the input portion of the matching network to
handle the comparatively high currents character-
istic of low impedance levels. Particular care was
taken in the design of the input part of the net-
work to reduce the normal resistive power losses
to a practical minimum.

To do this, the customary taps or links, with
their inherent resistive losses, were discarded in
favor of variable air-dielectric and fixed mica capaci-
tors, both of which have low loss characteristics.
The schematic diagram of this matching network
on page 113 is conventional except that the low-
value, high-impedance capacitor is divided into two
equal sections to provide a close—though not pre-
cise—balance with respect to ground, and to permit higher voltages. For low-power applications where the emphasis is on portability, the second tuning capacitor is omitted.

In either balanced or unbalanced form, the capacitive matching network is practical, versatile, and efficient. Matching losses can be cut to the point where they can be ignored in many instances. If a loop with a high ratio of inductance to capacitance (L/C ratio) is attempted, however, a significant portion of the circulating current flows through the stray (shunt) capacitance and a poor match and an inefficient coupling condition results. This could occur, for example, if the operating frequency is raised to a level closely approaching the self-resonant frequency formed by the loop inductor and the stray shunting capacitance.

In using the matching network, the transmitter is turned on. With the arrow on the pick-up probe in the "REFL" direction the meter reads the reverse power level. The controls are adjusted to the deepest null until the meter reads zero, indicating a perfect match. The transmitter power can then be advanced to a maximum level and the pick-up probe rotated to "FOR" to read the forward power level.

The method selected to reduce loss in the loop itself is a straightforward brute-force technique using conductors with an exceptionally large surface area. In the majority of the models built to date, the loop conductors are composed of 1½-inch tubing. With such large conductors, the r-f losses due primarily to skin effect are sharply pared.

**Radiation resistance**

The next step in improving efficiency is to obtain the greatest practical radiation resistance. This involves another classic equation:

\[ R_R = 3.12 \times 10^6 \left[ \frac{N^2}{A^2} \right] \lambda^2 \]

where \( R_R \) = the approximate radiation resistance of a small loop.

\( N = \) the number of turns in the loop.

\( A = \) the area enclosed.

\( \lambda = \) the wavelength of the frequency being used.

The dimensional terms are expressed in the same units, squared or linear.

The equation shows that two parameters, \( N \) and \( A \), may be controlled to possible advantage. In the case of \( N \), imagine a small loop antenna composed of a single turn shaped to form a square 3 feet on a side. Further assume that in all cases a very short drive cable is used, a low-to-moderate L/C ratio is employed, and losses in the capacitative matching network are negligible. In other words, essentially all the losses in this antenna stem from resistance in the wire conductor. Under these circumstances, the losses (\( R_R \)) are, to a great extent, directly proportional to the total length of the conductor.

As one would expect, the efficiency of this imaginary reference antenna is extremely low. In fact, if it were made of 18-gauge copper wire, the computed free-space efficiency at 3 Mhz would be a mere 0.0457%.

If the number of turns in a 3-foot-square loop were increased to three, the radiation resistance would be increased by a factor of nine. Unfortunately, the extra turns in this example also increase the loss resistance by a factor of three. Consequently, the efficiency improvement is reduced to a net factor of three; this is a reasonably significant improvement, but efficiency is still intolerably low.

The same conductor used for the three-turn loop can be reshaped to form a single-turn square loop with 9-foot sides, increasing its area by a factor of nine. Since \( R_R \) is proportional to \( A^2 \), the radiation resistance is now greater by a factor of 81. Here again, however, the losses are three times greater, resulting in a net efficiency improvement factor of 27. Using 18-gauge copper wire, the free-space calculated efficiency of this antenna is 1.23% at 3 Mhz and 13.1% at 6 Mhz.

Unless extremely severe restrictions are imposed on space, therefore, the largest practical single-turn loop is preferable to a smaller multturn unit.

**Shaping up**

The final consideration is form. Efficiency has been shown to be highest when a fixed-length peri-
Precise tuning. Reflected power meter permits fine tuning and exact matching, and monitors forward output power level. The coarse adjustments make discrete, step-by-step changes, while the fine controls provide continuous variations overlapping the coarse steps.

Meter encloses the greatest possible area. A circular configuration is best from a performance standpoint, but practical factors, such as the types of commercial tubing available, must be considered. This is especially true of developmental models, which are most economically built with straight tubing and "els."

Comparisons were made between a square, an equilateral octagon, and a circle, all of the same total perimeter. The octagon encloses 20% more space than the square, and the circle 29% more. Since radiation resistance is proportional to the square of the area, the radiation efficiency of the small octagon loop exceeds that of the square by about 45%, while the circular loop outperforms its square counterpart by about 65%. But because straight tubing packs more compactly, is more readily available, and is much cheaper than the circular kind, the octagon shape was chosen. The size—5 feet is the maximum length for any part—was held down so the antenna could be easily disassembled and carried in a small motor vehicle.

With the loss reduction achieved, the loop antennas that have been built provide a calculated free-space efficiency of about 22% at 2.5 Mhz and about 77% at 5 Mhz.

In the final analysis, however, an antenna's efficiency, or, perhaps more important, its effectiveness, must be proven in field tests.

As noted before, preliminary tests in the U.S. were followed by several hundred more conducted by the Army in Vietnam. Numerous frequencies and different ranges were used at various times of day in the later trials. Also, the tests were at all times confined to relatively low power levels, usually on the order of a few watts.

Nevertheless, all contacts attempted in Vietnam were not only 100% successful, but, with few exceptions, were rated "perfect." These few exceptions were never ranked lower than "good."

**Competitive trials**

The earlier U.S. tests, however, were perhaps more informative in providing direct comparisons between the loop and other antennas. As a matter of fact, the octagonal loop made of ½-inch tubing is actually a second-generation model. For convenient construction, the earlier developmental models were constructed in a square shape and formed from ¼-inch tubing. Later tests have revealed a very significant 8-decibel superiority for the octagon version over the square.

Despite this, however, the square models showed themselves to be effective radiators in the early trials. In numerous ionospheric field tests conducted between Aberdeen Proving Ground and two widely separated mountain valleys in the Alleghenies, the square loop proved, in almost all instances, as effective as a full-sized dipole, and actually surpassed the dipole in most matchups. Several frequencies between 2.5 Mhz and 5.5 Mhz were used, and in each instance the dipole was cut to the optimum length. Furthermore, the dipole was strung in a cleared area at a height of 40 feet, while the bottom of the loop was only 4 feet above the ground.

In one field test to determine ground-wave propagation over a distance of two miles, a dipole supported at a height of 6 feet delivered a measured field strength of about 68 microvolts on a transmitter power of 6 watts. The square loop gave about 600 microvolts on the same power, and the octagon produced approximately 1,500 microvolts. The frequency used was 3.275 Mhz—not the best for the small loops.

It's true that the dipole's propagation in this particular test suffered from a less favorable polarization. But polarization characteristics are basically inherent and have to be considered in choosing antennas for operation at low heights.

**Modifications**

Though the octagon loop is a second-generation version, further improvements are being made. With the fiberglass mast used in some recent models, a single soldier can set up the antenna. For some applications, the diameter of the tubing can perhaps be advantageously increased from the present ½ inches to 3, 4, or even to 6 inches. For other applications, the usual output tank circuit and matching network can be eliminated if the final amplifier is located close to the loop. The loop then can serve as both the radiation element and the final tank circuit, a technique especially useful where power is low and portability a prime requisite. Models of such long-range, highly portable antennas have been successfully tested at the Limited War Laboratory.